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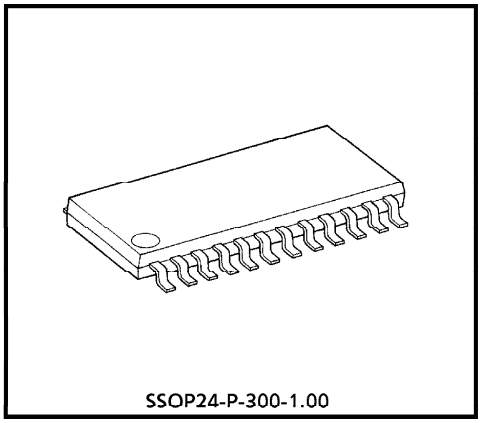
TOSHIBA Bi-CMOS INTEGRATED CIRCUIT SILICON MONOLITHIC

TB6504F

PWM CHOPPER TYPE BIPOLAR STEPPING MOTOR DRIVER.

The TB6504F is PWM chopper type sinusoidal micro step bipolar stepping motor driver.
Sinusoidal micro step operation is accomplished only a clock signal inputting by means of built-in hard ware.



SSOP24-P-300-1.00

Weight : 0.32g (Typ.)

FEATURES

- 1 chip bipolar sinusoidal micro step stepping motor driver.
- Output Current up to 150mA
- PWM chopper type.
- Structured by high voltage Bi-CMOS process technology.
- Forward and reverse rotation are available.
- 2, 1-2, W1-2, 2W1-2 phase 1 or 2 clock drives are selectable.
- Package : SSOP24-P-300-1.00
- Input Pull-Up Resistor equipped with RESET and ENABLE Terminal : R = 200kΩ (Typ.)
- Output Monitor available with $\overline{MO}$. $I_{O\,(\overline{MO})}$ = ±2mA MAX.
- Reset and Enable are available with $\overline{RESET}$ and $\overline{ENABLE}$.

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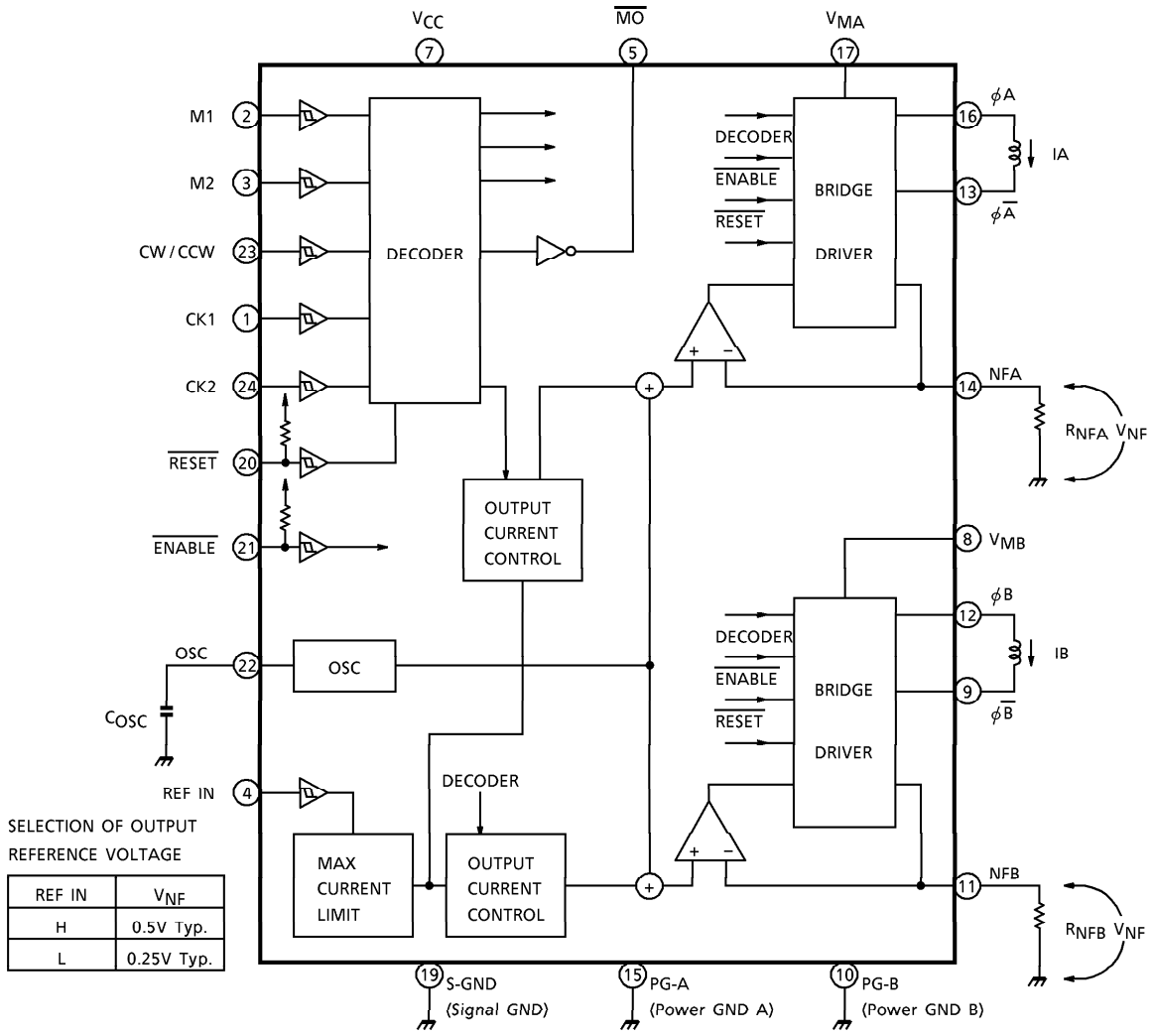
BLOCK DIAGRAM



SELECTION OF OUTPUT REFERENCE VOLTAGE

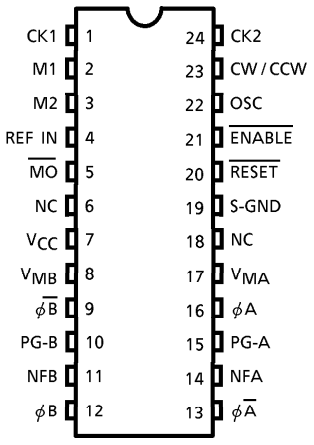
REF IN	V_{NF}
H	0.5V Typ.
L	0.25V Typ.

Pull-up Resistance pin⑳, ㉑ : 200kΩ (Typ.)
pin⑥, ⑱ : Non Connection

PIN FUNCTION

PIN No.	SYMBOL	FUNCTIONAL DESCRIPTION	
1	CK1	Clock signal input terminal.	TRUTH TABLE A
2	M1	Excitation control input.	TRUTH TABLE B
3	M2	Excitation control input.	
4	REF IN	V_{NF} control input. High Level ; V_{NF} = 0.5V, Low Level ; V_{NF} = 0.25V	
5	$\overline{MO}$	Monitor output.	
6	NC	No connection.	
7	V_{CC}	Supply voltage terminal for contol circuit.	
8	V_{MB}	Supply voltage terminal for Motor Drive.	
9	$\phi\overline{B}$	Output $\overline{B}$	
10	PG-B	Power GND	
11	NFB	B-ch current detection terminal.	
12	ϕB	Output B	
13	$\phi\overline{A}$	Output $\overline{A}$	
14	NFA	A-ch current detection terminal.	
15	PG-A	Power GND	
16	ϕA	Output A.	
17	V_{MA}	Supply voltage terminal for Motor Drive.	
18	NC	No connection.	
19	S-GND	Signal GND.	
20	$\overline{RESET}$	Reset signal input terminal.	TRUTH TABLE A
21	$\overline{ENABLE}$	Enable signal input terminal.	
22	OSC	Sawtooth oscilation terminal.	
23	CW / CCW	Forward rotation / Reverse rotation input terminal.	TRUTH TABLE A
24	CK2	Clock signal input terminal.	

PIN CONNECTION (Top view)



(Note) NC : No connection

TRUTH TABLE A

INPUT					MODE
CK1	CK2	CW / CCW	$\overline{RESET}$	$\overline{ENABLE}$	
↑	H	L	H	L	CW
⊓	L	L	H	L	INHIBIT
H	↑	L	H	L	CCW
L	⊓	L	H	L	INHIBIT
↑	H	H	H	L	CCW
⊓	L	H	H	L	INHIBIT
H	↑	H	H	L	CW
L	⊓	H	H	L	INHIBIT
X	X	X	L	L	INITIAL
X	X	X	X	H	Z

Z : High impedance
X : Don't Care

TRUTH TABLE B

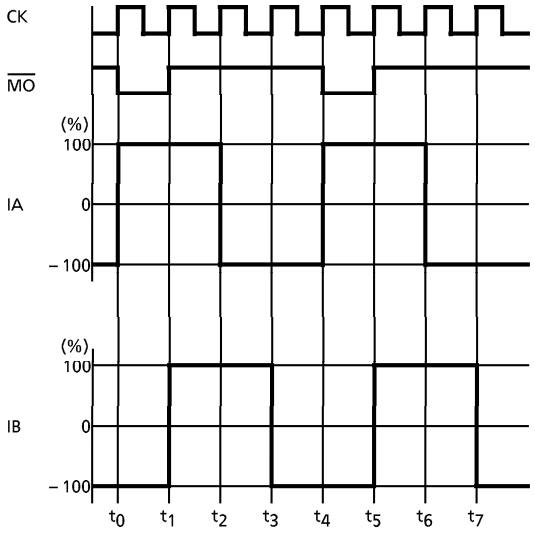
INPUT		MODE (EXCITATION)
M1	M2	
L	L	2 Phase
H	L	1-2 Phase
L	H	W1-2 Phase
H	H	2W1-2 Phase

INITIAL MODE

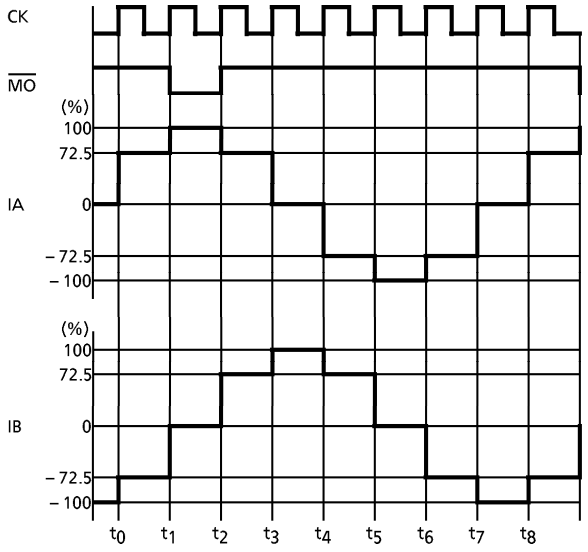
MODE	I_{OUT} (A)	I_{OUT} (B)
2 Phase	100%	− 100%
1-2 Phase	100%	0%
W1-2 Phase	100%	0%
2W1-2 Phase	100%	0%

EXCITATION

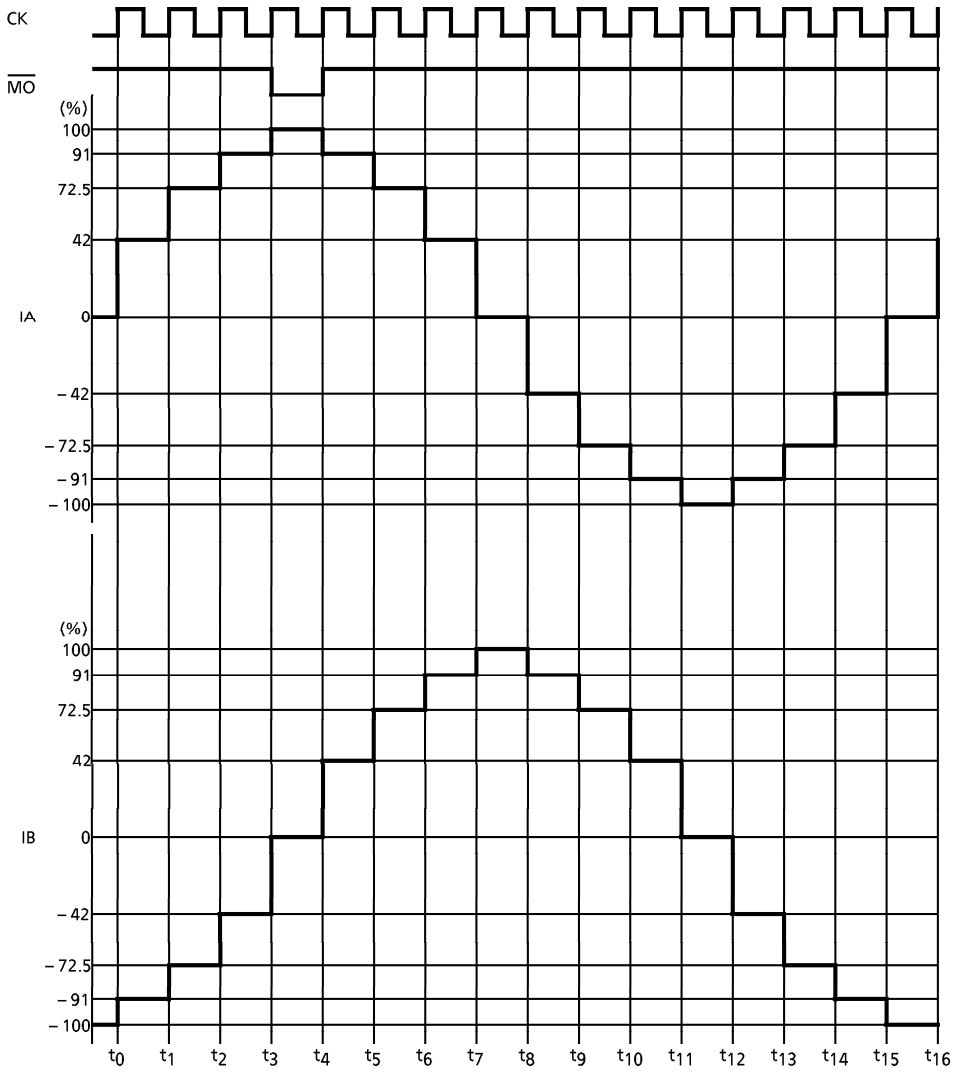
2 Phase excitation (M1 : L, M2 : L, CW MODE)



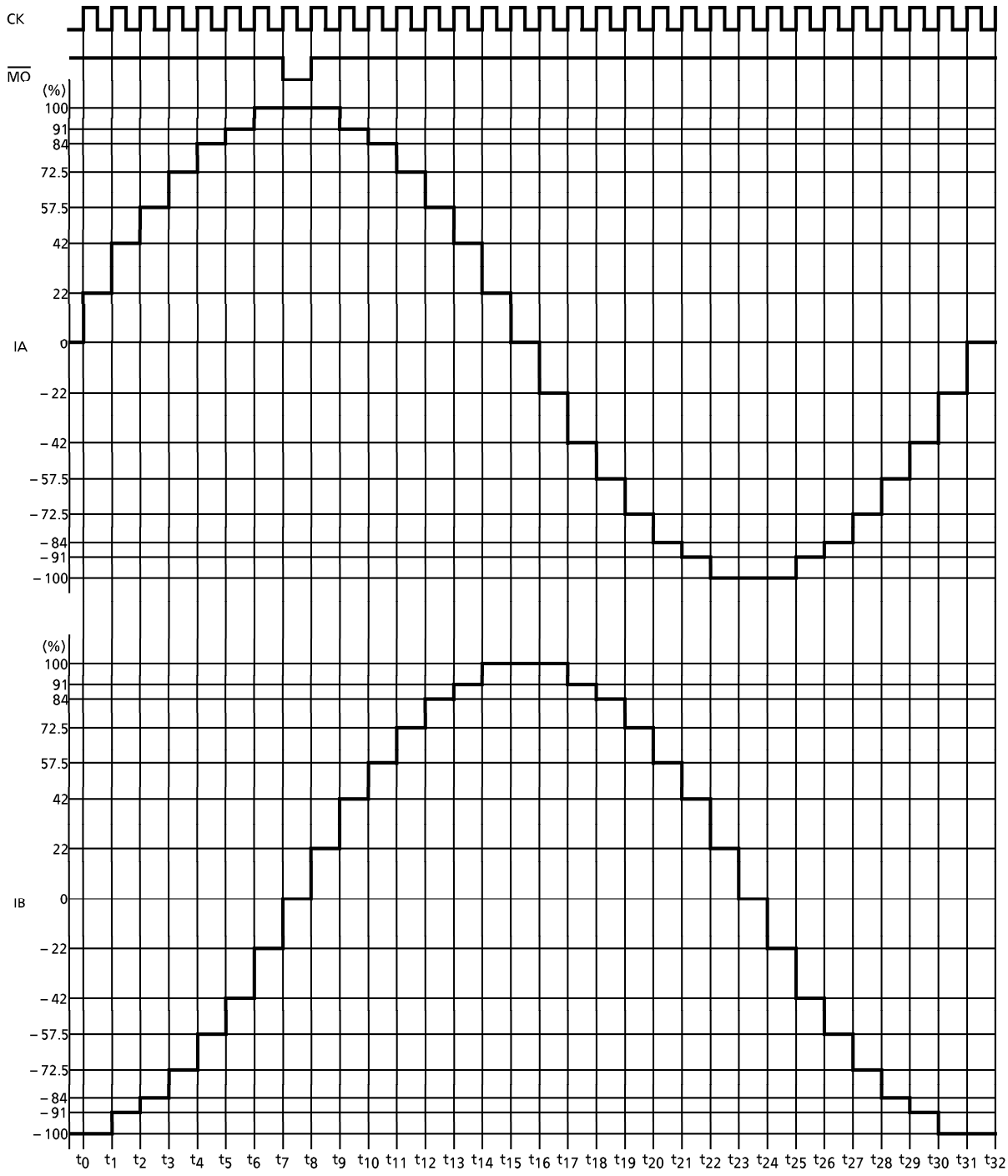
1-2 Phase excitation (M1 : H, M2 : L, CW MODE)



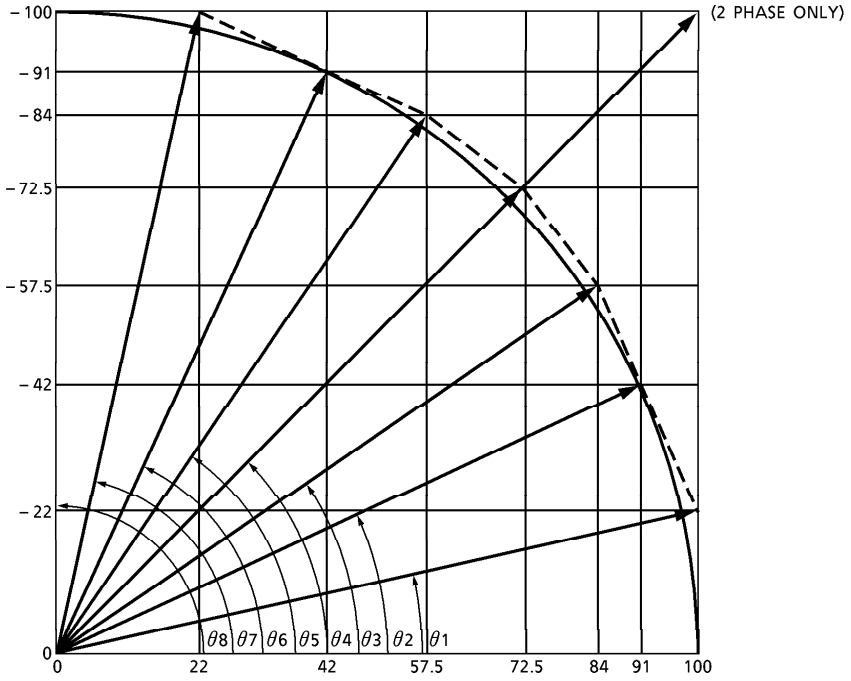
W1-2 Phase excitation (M1 : L, M2 : H, CW MODE)



2W1-2 Phase excitation (M1 : H, M2 : H, CW MODE)

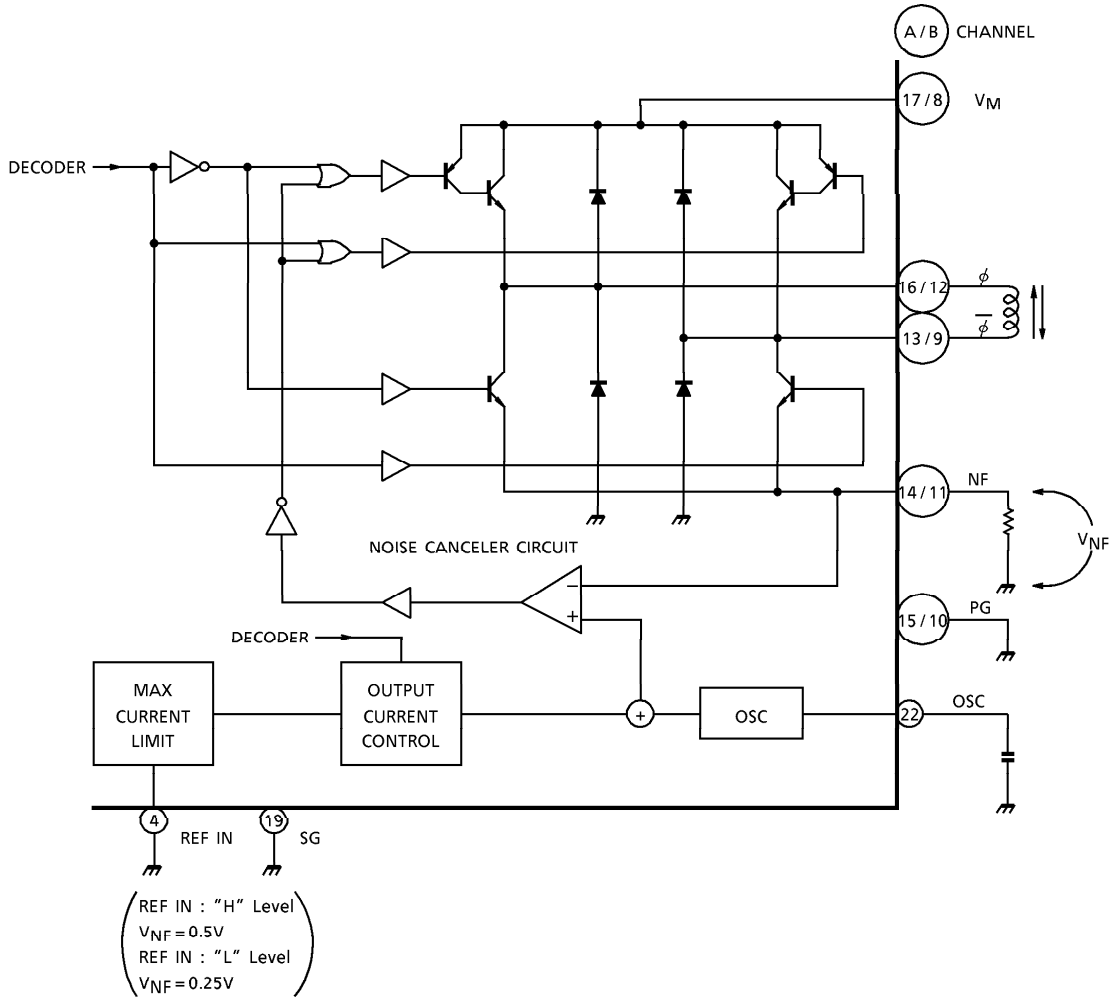


OUTPUT CURRENT VECTOR ORBIT (Normalize to 90° for each one step)



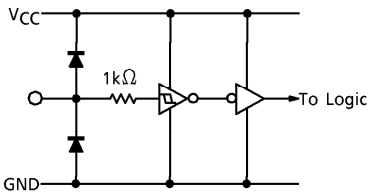
θ	ROTATION ANGLE		VECTOR LENGTH		
	IDEAL	TB6504F	IDEAL	TB6504F	
θ0	0°	0°	100	100.00	—
θ1	11.25°	12.41°	100	102.39	—
θ2	22.5°	27.78°	100	100.22	—
θ3	33.75°	34.39°	100	101.80	—
θ4	45°	45°	100	102.53	141.42
θ5	56.25°	55.61°	100	101.81	—
θ6	67.5°	65.22°	100	100.22	—
θ7	78.75°	77.59°	100	102.39	—
θ8	90°	90°	100	100.00	—
			1-2, W1-2, 2W1-2 Phase		2 Phase

OUTPUT CIRCUIT

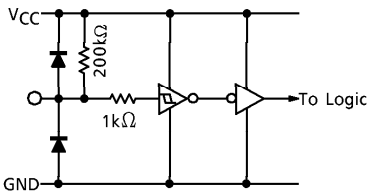


INPUT CIRCUIT

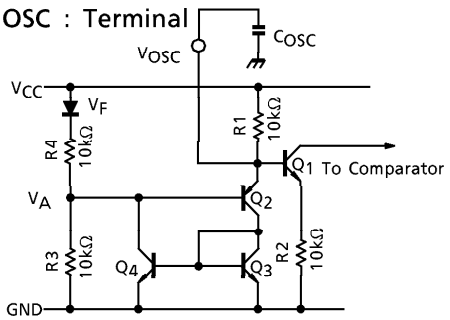
CK1, CK2, CW / CCW, M1, M2, REF IN : Terminals



$\overline{\text{RESET}}$, $\overline{\text{ENABLE}}$: Terminal



OSC : Terminal



$$V_A = (V_{CC} - V_F)\frac{R3}{R3 + R4} \fallingdotseq 2.15V$$

OSC FREQUENCY CALCULATION

Sawtooth OSC circuit consists of Q_1 through Q_4 and R1 through R4.
Q_2 is turned "off" when V_{OSC} is less than the voltage of $2.5V + V_{BE}$ Q_2 approximately equal to 2.85V.
V_{OSC} is increased by C_{OSC} charging through R1.
Q_3 and $Q4$ are turned "on" when V_{OSC} becomes 2.85V (Higher level.)
Lower level of V ㉒ pin is equal to V_{BE} $Q_2 + V_{SAT}$ Q_4 approximately equal to 1.4V.
V_{OSC} is calculated by following equation.

$$V_{OSC} = 5 \cdot [1 - \exp(-\frac{t}{C_{OSC} \cdot R1})] \quad \cdots\cdots\cdots\cdots\cdots\cdots\cdots \quad ①$$

Assuming that $V_{OSC} = 1.4V$ $(t = t_1)$ and $= 2.85V$ $(t = t_2)$
C_{OSC} is external capacitance connected to pin㉒ and R1 is on-chip 10kΩ resistor.
Therefore, OSC frequency is calculated as follows.

$$t_1 = -C_{OSC} \cdot R1 \cdot \ell n (1 - \frac{1.4}{5}) \quad \cdots\cdots\cdots\cdots\cdots\cdots\cdots \quad ②$$

$$t_2 = -C_{OSC} \cdot R1 \cdot \ell n (1 - \frac{2.85}{5}) \quad \cdots\cdots\cdots\cdots\cdots\cdots\cdots \quad ③$$

$$f_{OSC} = \frac{1}{t_2 - t_1} = \frac{1}{C_{OSC}(R1 \cdot \ell n (1 - \frac{1.4}{5}) - R1 \cdot \ell n (1 - \frac{2.85}{5}))}$$

$$= \frac{1}{5.15 \cdot C_{OSC}} (kHz) \; (C_{OSC} : \mu F)$$

$\overline{\text{ENABLE}}$ AND $\overline{\text{RESET}}$ FUNCTION AND $\overline{\text{MO}}$ SIGNAL

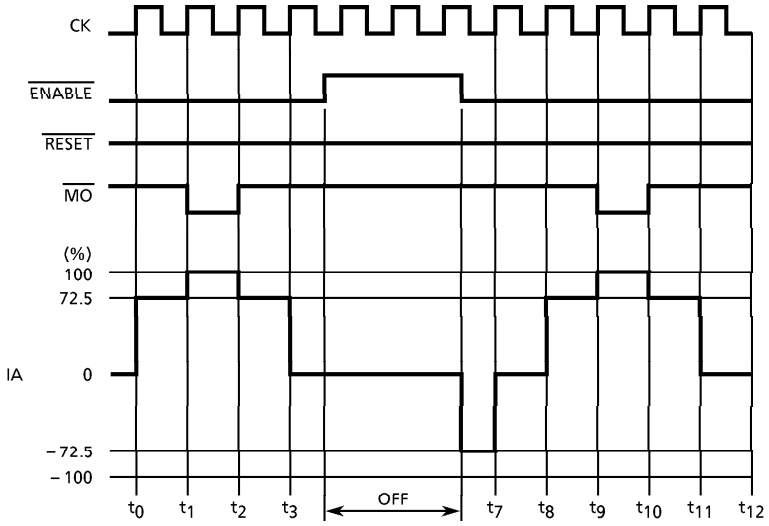


Fig.1 1-2 Phase drive mode (M1 : H, M2 : L)

$\overline{\text{ENABLE}}$ Signal disables only Output Signal.
Internal logic functions are proceeded by CK signal without regard to $\overline{\text{ENABLE}}$ signal.
Therefore, Output Current is initiated from the proceeded timing point of internal logic circuit after release of disable mode.
Fig.1 shows the $\overline{\text{ENABLE}}$ functions, when the system is selected in 1-2 Phase drive mode.
As $\overline{\text{RESET}}$ is low, the decoder is initialized and $\overline{\text{MO}}$ is low.
After $\overline{\text{RESET}}$ is high, the motion is resumed from next clock as shown in Fig.2.

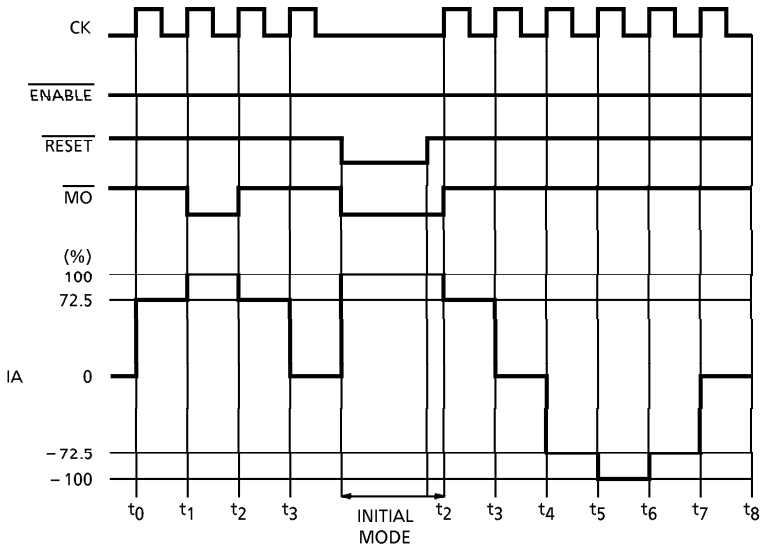


Fig.2 1-2 Phase drive mode (M1 : H, M2 : L)

$\overline{\text{MO}}$ (Monitor Output) Signals is used as rotaion and initial signal for stable rotation checking.

MAXIMUM RATINGS (Ta = 25°C)

CHARACTERISTIC	SYMBOL	RATING	UNIT
Supply Voltage	V_{CC}	5.5	V
	$V_{M\ (opr)}$	V_{CC} – 0.3~10	
	$V_{M\ (MAX)}$	18	
Output Current	$I_{O\ (MAX)}$	150	mA
	$I_{O\ (\overline{MO})}$	± 2	
Input Voltage	V_{IN}	~V_{CC}	V
Power Dissipation	P_D	0.59 (Note 1)	W
		0.83 (Note 2)	
Operating Temperature	T_{opr}	– 10~70	°C
Storage Temperature	T_{stg}	– 55~150	°C
Feed Back Voltage	V_I	1.0	V

(Note 1) No heat sink
(Note 2) With heat sink (50 × 50 × 1.6mm Cu 10%)

RECOMMENDED OPERATING CONDITIONS (Ta = – 10~70°C)

CHARACTERISTIC	SYMBOL	TEST CIR-CUIT	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Supply Voltage	$V_{CC\ (opr)}$	—	—	4.5	5.0	5.5	V
Output Voltage	$V_{M\ (opr)}$	—	—	5.5	—	8.0	V
Output Current	I_{OUT}	—	—	—	—	120	mA
Input Voltage	V_{IN}	—	—	—	—	V_{CC}	V
Clock Frequency	f_{CLOCK}	—	—	—	—	5	kHz
OSC Frequency	f_{OSC}	—	—	15	—	80	kHz

ELECTRICAL CHARACTERISTICS (Unless otherwise specified, Ta = 25°C, V_{CC} = 5V, V_M = 8V)

CHARACTERISTIC		SYMBOL	TEST CIR-CUIT	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Input Voltage	High	$V_{IN\ (H)}$	1	M1, M2, CW / CCW, REF IN $\overline{ENABLE}$, CK1, CK2, $\overline{RESET}$	3.5	—	V_{CC} + 0.4	V
	Low	$V_{IN\ (L)}$			GND − 0.4	—	1.5	
Input Hysteresis Voltage		V_H			—	600	—	mV
Input Current		$I_{IN\text{-}1\ (H)}$	1	M1, M2, REF IN, V_{IN} = 5.0V	—	—	100	nA
		$I_{IN\text{-}1\ (L)}$		$\overline{RESET}$, V_{IN} = 0V, $\overline{ENABLE}$ INTERNAL PULL-UP RESISTOR	5	25	50	μA
		$I_{IN\text{-}2\ (L)}$		SOURCE TYPE, V_{IN} = 0V	—	—	100	nA
Quiescent Current V_{CC} Terminal		I_{CC1}	2	Output Open $\overline{RESET}$: H $\overline{ENABLE}$: L (2, 1-2 Phase excitation)	—	10	18	mA
		I_{CC2}		Output Open (W1-2, 2W1-2 Phase excitation) $\overline{RESET}$: H $\overline{ENABLE}$: L	—	10	18	
		I_{CC3}		$\overline{RESET}$: L, $\overline{ENABLE}$: H	—	5	—	
		I_{CC4}		$\overline{RESET}$: H, $\overline{ENABLE}$: H	—	5	—	
Comparator Reference Voltage	High	$V_{NF\ (H)}$	3	REF IN H R_{NF} = 5Ω, C_{OSC} = 0.0033μF	0.45	0.5	0.55	V
	Low	$V_{NF\ (L)}$		REF IN L R_{NF} = 2.5Ω, C_{OSC} = 0.0033μF	0.22	0.25	0.28	
Output Differential		ΔV_O	—	B / A, C_{OSC} = 0.0033μF R_{NF} = 2.5Ω, REF IN = L	− 10	—	10	%
$V_{NP\ (H)} - V_{NF\ (L)}$		ΔV_{NF}	—	$V_{NF\ (L)} / V_{NF\ (H)}$ C_{OSC} = 0.0033μF	43	50	57	%
Maximum OSC Frequency		$f_{OSC\ (MAX.)}$	—	—	100	—	—	kHz
Minimum OSC Frequency		$f_{OSC\ (MIN.)}$	—	—	—	—	10	kHz
OSC Frequency		f_{OSC}	—	C_{OSC} = 0.0033μF	31	44	70	kHz
Output Voltage		$V_{OH\ (MO)}$	—	I_{OH} = − 40μA	4.5	4.9	V_{CC}	V
		$V_{OL\ (MO)}$	—	I_{OL} = 40μA	GND	0.1	0.5	

OUTPUT BLOCK

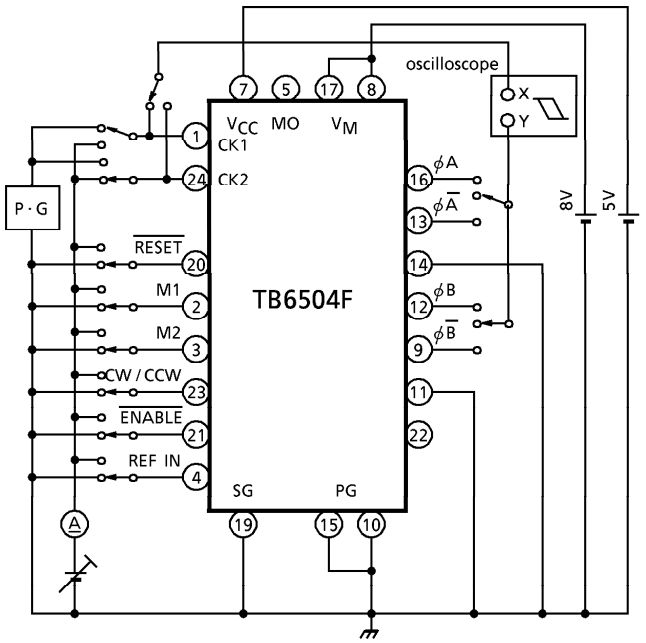
CHARACTERISTIC				SYMBOL	TEST CIR-CUIT	TEST CONDITION		MIN.	TYP.	MAX.	UNIT
Output Saturation Voltage	Upper Side			$V_{SAT\ U1}$	4	I_{OUT} = 0.12A		—	0.90	1.25	V
	Lower Side			$V_{SAT\ L1}$				—	0.22	0.37	
	Upper Side			$V_{SAT\ U2}$		I_{OUT} = 0.06A		—	0.83	—	
	Lower Side			$V_{SAT\ L2}$				—	0.12	—	
Diode Forward Voltage	Upper Side			$V_{F\ U1}$	5	I_{OUT} = 0.12A		—	1.18	1.8	V
	Lower Side			$V_{F\ L1}$				—	0.92	1.6	
Output Dark Current (A + B Channels)				I_{M1}	2	$\overline{ENABLE}$: "H" Level $\overline{RESET}$: "L" Level Output Open		—	—	50	μA
				I_{M2}		$\overline{ENABLE}$: "L" Level $\overline{RESET}$: "H" Level Output Open, 2 Phase excitation mode		—	8	28	mA
NF Terminal Current				I_{NF}		$\overline{ENABLE}$: "L" Level $\overline{RESET}$: "H" Level Output Open		1	2.5	7	
A-B Chopping Current (Note)	2W1-2φ	W1-2φ	1-2φ	VECTOR	3	θ = 0	REF IN : L RNF = 2.5Ω C_{OSC} = 0.0033μF L = 10mH / R = 0.5Ω	—	100	—	%
	2W1-2φ	—	—			θ = 1 / 8		—	100	—	
	2W1-2φ	W1-2φ	—			θ = 2 / 8		86	91	96	
	2W1-2φ	—	—			θ = 3 / 8		79	84	89	
	2W1-2φ	W1-2φ	1-2φ			θ = 4 / 8		67.5	72.5	77.5	
	2W1-2φ	—	—			θ = 5 / 8		52.5	57.5	62.5	
	2W1-2φ	W1-2φ	—			θ = 6 / 8		37	42	47	
	2W1-2φ	—	—			θ = 7 / 8		17	22	27	
	2 Phase Excitation Mode VECTOR			—	—	—		—	100	—	

(Note) Maximum current (θ = 0) : 100%
2W1-2φ : 2W1, 2 phase excitation mode
W1-2φ : W1, 2 phase excitation mode
1-2φ : 1, 2 phase excitation mode

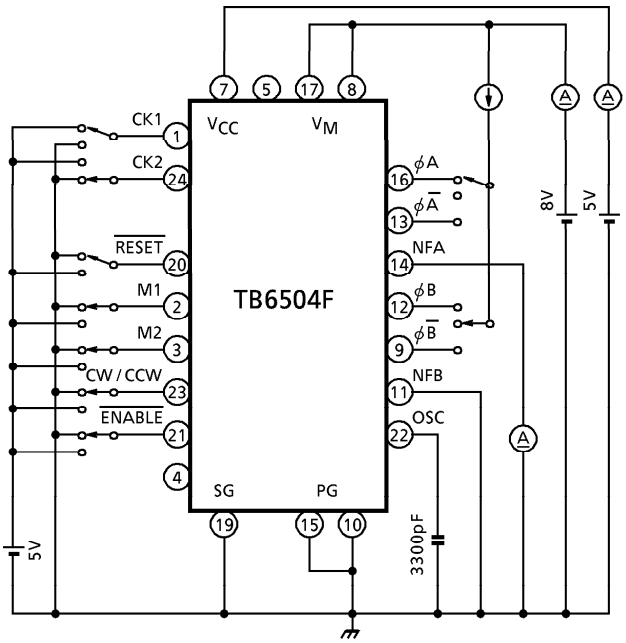
CHARACTERISTIC				SYMBOL	TEST CIR-CUIT	TEST CONDITION		MIN.	TYP.	MAX.	UNIT
A-B Chopping Current (Note)	2W1-2ϕ	W1-2ϕ	1-2ϕ	VECTOR	3	θ = 0	REF IN : L RNF = 3.3Ω C_{OSC} = 0.0033μF L = 20mH / R = 60Ω	—	100	—	%
	2W1-2ϕ	—	—			θ = 1 / 8		—	100	—	
	2W1-2ϕ	W1-2ϕ	—			θ = 2 / 8		—	91.2	—	
	2W1-2ϕ	—	—			θ = 3 / 8		—	84.2	—	
	2W1-2ϕ	W1-2ϕ	1-2ϕ			θ = 4 / 8		—	73.6	—	
	2W1-2ϕ	—	—			θ = 5 / 8		—	59	—	
	2W1-2ϕ	W1-2ϕ	—			θ = 6 / 8		—	44.6	—	
	2W1-2ϕ	—	—			θ = 7 / 8		—	25.6	—	
	2 Phase Excitation Mode VECTOR					—		—	100	—	
Feed Back Voltage Step				ΔV_{NF}	—	$\Delta\theta$ = 0 / 8 – 1 / 8	REF IN : L RNF = 2.5Ω C_{OSC} = 0.0033μF	—	0	—	mV
						$\Delta\theta$ = 1 / 8 – 2 / 8		10	22.5	35	
						$\Delta\theta$ = 2 / 8 – 3 / 8		5	17.5	30	
						$\Delta\theta$ = 3 / 8 – 4 / 8		16.25	28.75	41.25	
						$\Delta\theta$ = 4 / 8 – 5 / 8		25	37.5	50	
						$\Delta\theta$ = 5 / 8 – 6 / 8		26.25	38.75	51.25	
						$\Delta\theta$ = 6 / 8 – 7 / 8		37.5	50	62.5	
Output Tr Switching Characteristics				t_r	7	R_L = 2Ω, V_{NF} = 0V, C_L = 15pF		—	0.3	—	μs
				t_f				—	2.2	—	
				t_{pLH}		CK~Output		—	1.5	—	
				t_{pHL}				—	2.7	—	
				t_{pLH}		OSC~Output		—	5.4	—	
				t_{pHL}				—	6.3	—	
				t_{pLH}		$\overline{RESET}$~Output		—	2.0	—	
				t_{pHL}				—	2.5	—	
				t_{pLH}		$\overline{ENABLE}$~Output		—	5.0	—	
				t_{pHL}				—	6.0	—	
Output Leakage Current	Upper Side			I_{OH}	6	V_M = 18V		—	—	50	μA
	Lower Side			I_{OL}				—	—	50	

(Note) Maximum current (θ = 0) : 100%
2W1-2ϕ : 2W1, 2 phase excitation mode
W1-2ϕ : W1, 2 phase excitation mode
1-2ϕ : 1, 2 phase excitation mode

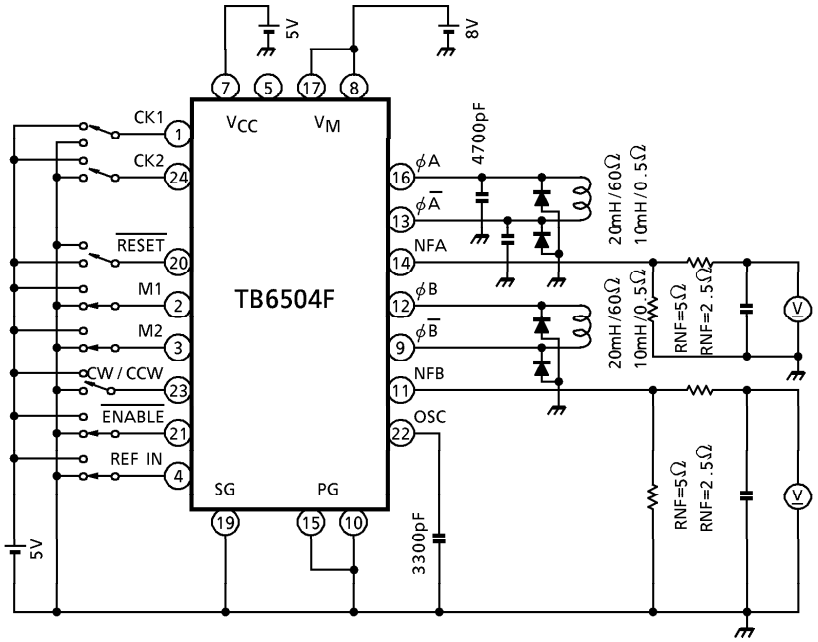
TEST CIRCUIT 1. : V_{IN} (H), (L), I_{IN} (H), (L)



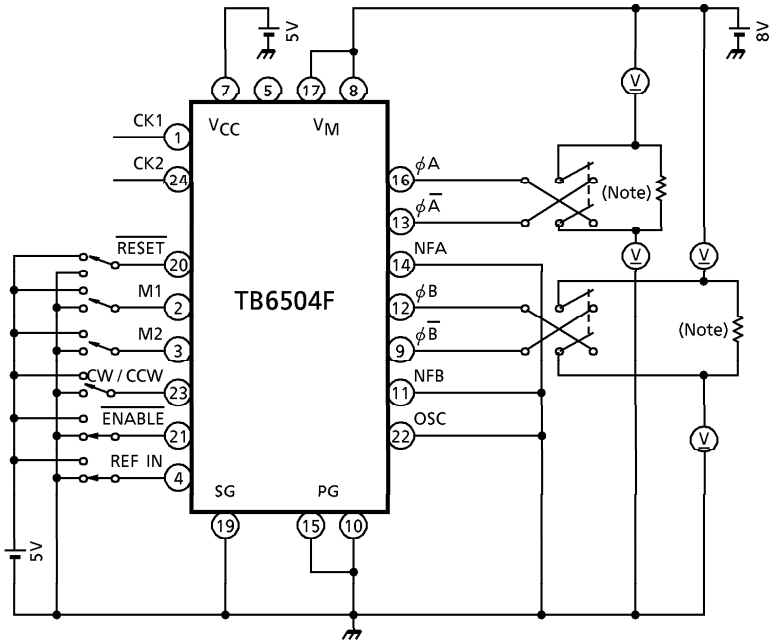
TEST CIRCUIT 2. : I_{CC}, I_M, I_{NF}



TEST CIRCUIT 3. : V_{NF} (H), (L)

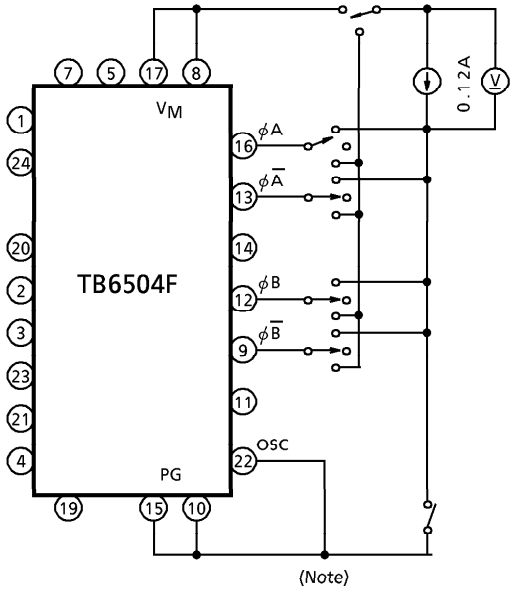


TEST CIRCUIT 4. : $V_{CE\ (SAT)}$ Upper, Lower

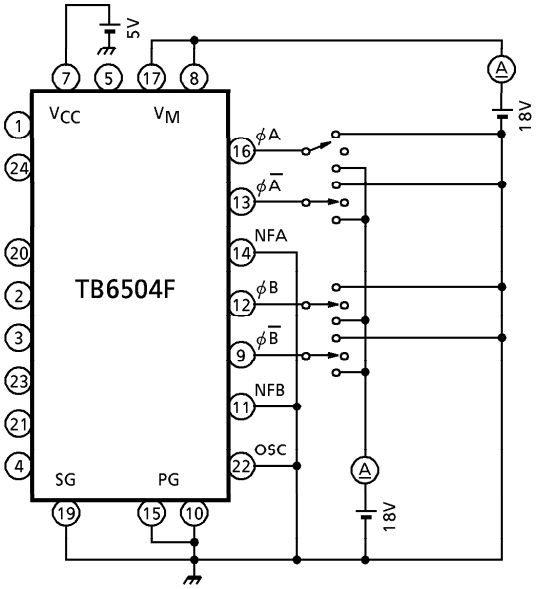


(Note) Calibrate Output Current becomes 0.06A (or 0.12A) with this resistor.

TEST CIRCUIT 5. : V_{F-U}, V_{F-L}



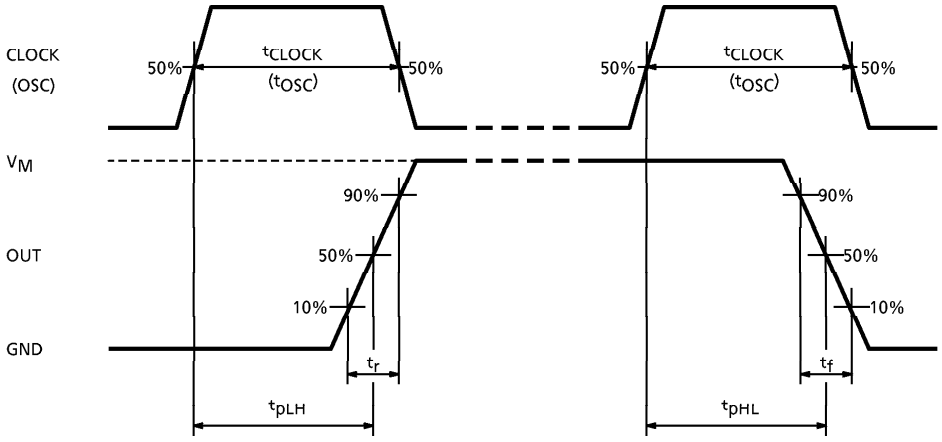
TEST CIRCUIT 6. : I_{OH}, I_{OL}

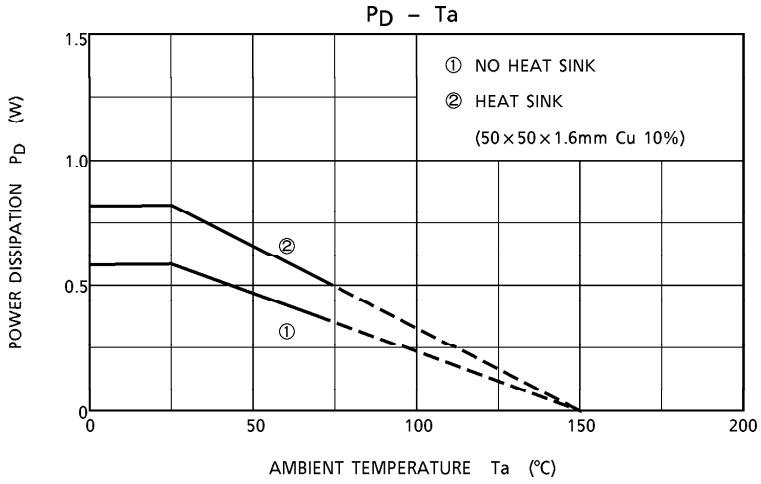


(Note) Not to take a GND with any non-connecting Pins.

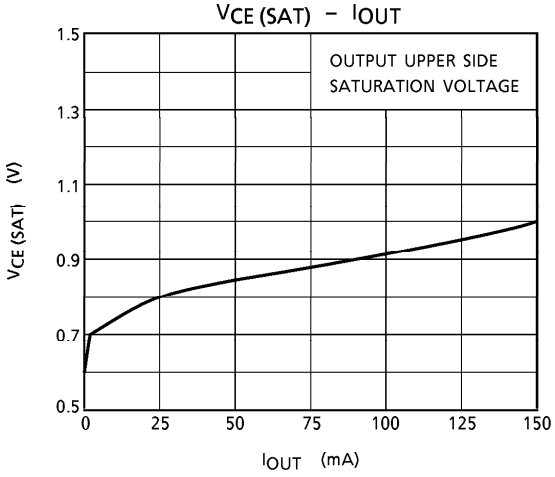
AC ELECTRICAL CHARACTERISTIC, TEST CIRCUIT

CK (OSC)-OUT

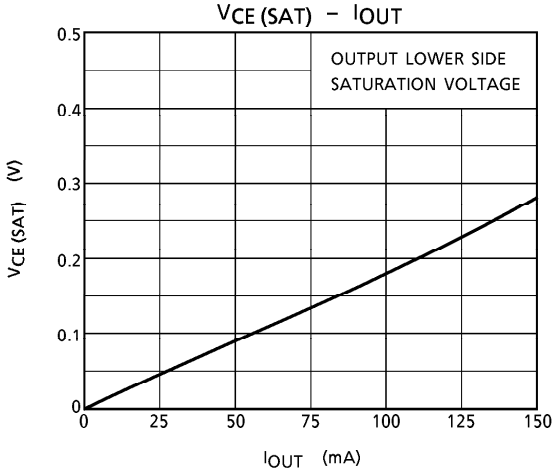




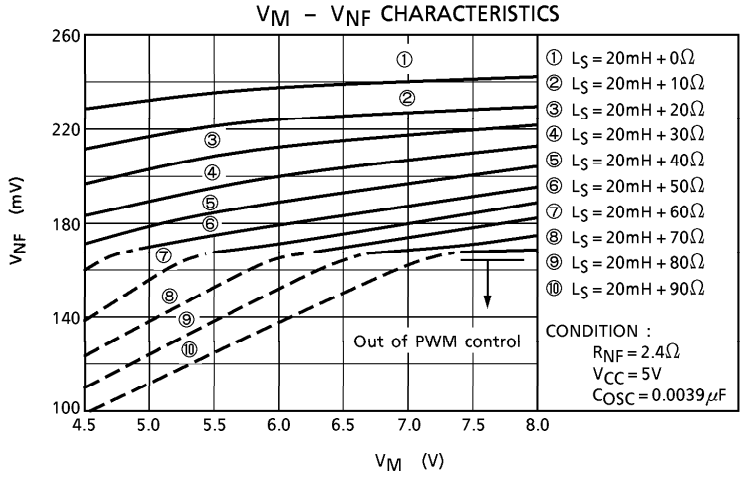
$V_{CE\,(SAT)}$ – I_{OUT}



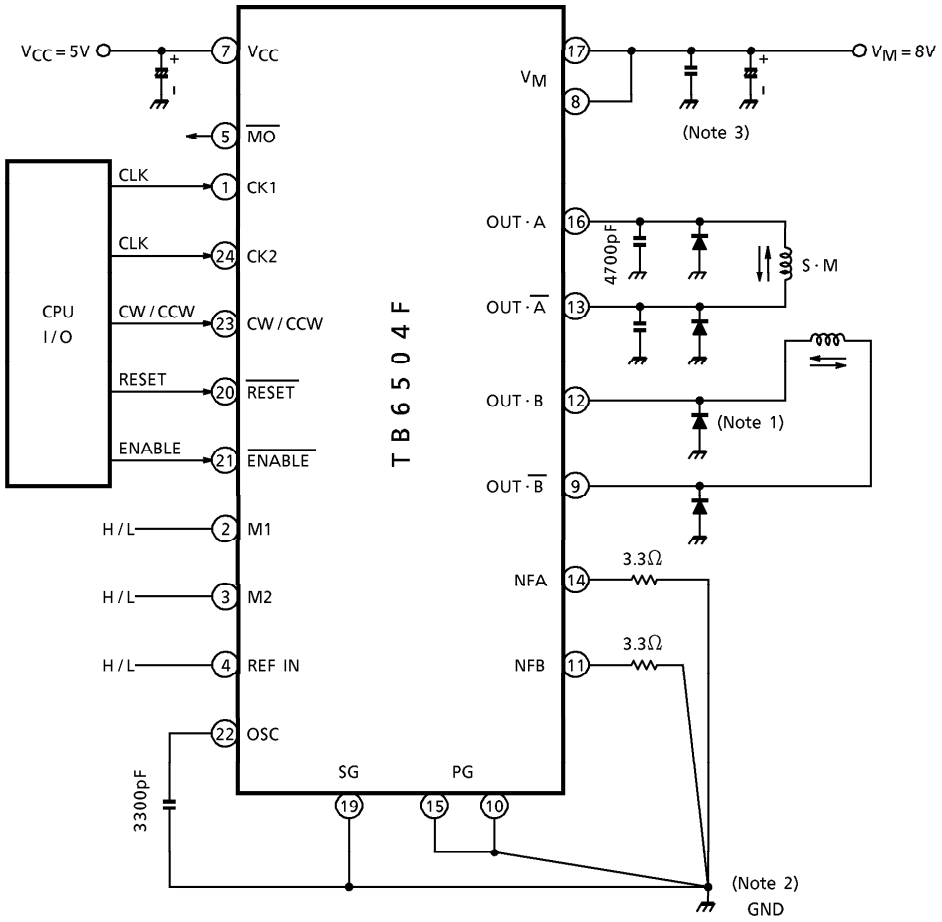
$V_{CE\,(SAT)}$ – I_{OUT}



V_M – V_{NF} CHARACTERISTICS



APPLICATION CIRCUIT



(Note 1) Schottky diode (U1GWJ49) to be connected additionally between each output (pin 16 / 13 / 12 / 9) and GND for preventing Punch-through Current.

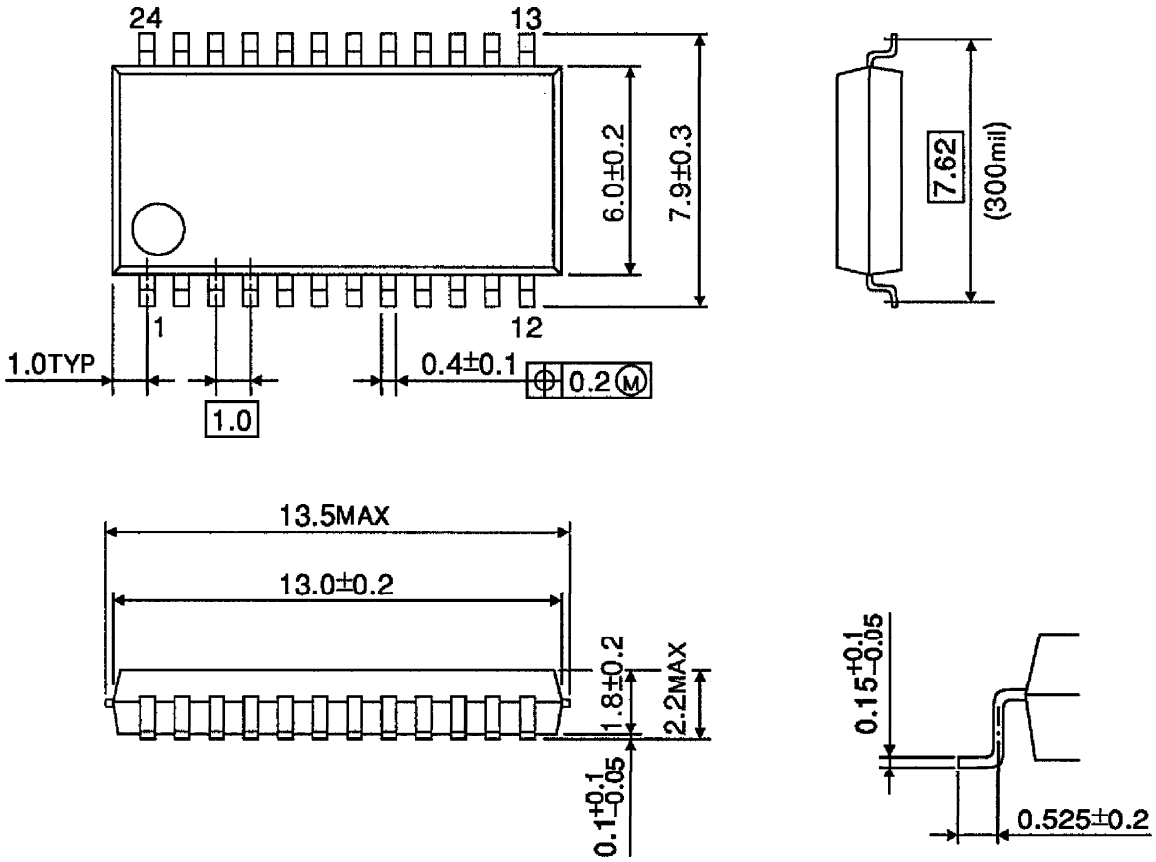
(Note 2) GND pattern to be laid out at one point in order to prevent common impedance.

(Note 3) Capasitor for noise suppression to be connected between the Power Supply (V_{CC}, V_M) and GND to stabilize the operation.

(Note 4) Utmost care is necessary in the design of the output line, V_M and GND line since IC may be destroyed due to short-circuit between outputs, air contamination fault, or fault by improper grounding.

OUTLINE DRAWING

SSOP24-P-300-1.00 Unit : mm



Weight : 0.32g (Typ.)